



DR. GYURCSEK ISTVÁN

# Exercises with Two-Ports

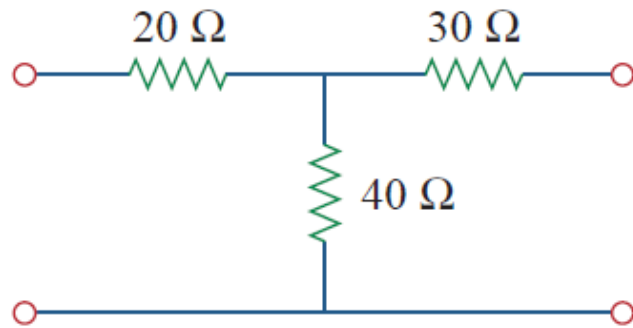
*Sources and additional materials (recommended)*

- ❑ *Dr. Gyurcsek – Dr. Elmer: Theories in Electric Circuits, GlobeEdit, 2016, ISBN:978-3-330-71341-3*
- ❑ *Ch. Alexander, M. Sadiku: Fundamentals of Electric Circuits, 6th Ed., McGraw Hill NY 2016, ISBN: 978-0078028229*
- ❑ *Dr. Selmeczi K. – Schnöller A.: Villamosságtan 1. MK Budapest 2002, TK szám: 49203/I*
- ❑ *Dr. Selmeczi K. – Schnöller A.: Villamosságtan 2. TK Budapest 2002, ISBN:9631026043*

# Z Parameters (OC)

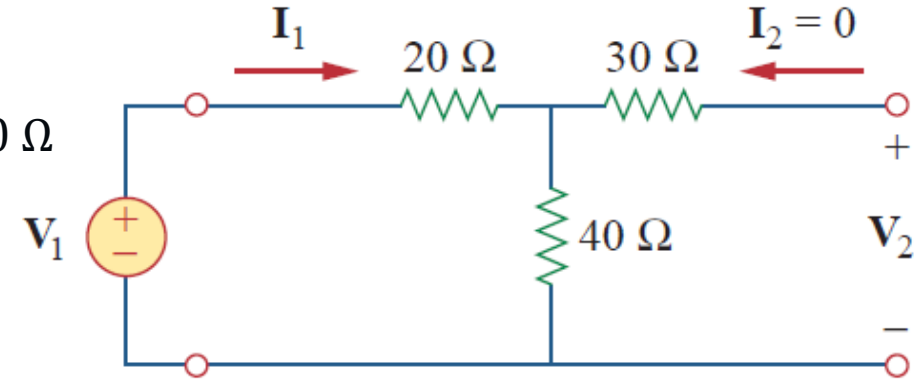


Example **TWO.01** – Determine the Z parameters of the T-section circuit.



$$z_{11} = \left. \frac{V_1}{I_1} \right|_{I_2=0} = \frac{I_1 \cdot (20 + 40)}{I_1} = 60 \Omega$$

$$z_{21} = \left. \frac{V_2}{I_1} \right|_{I_2=0} = \frac{I_1 \cdot 40}{I_1} = 40 \Omega$$



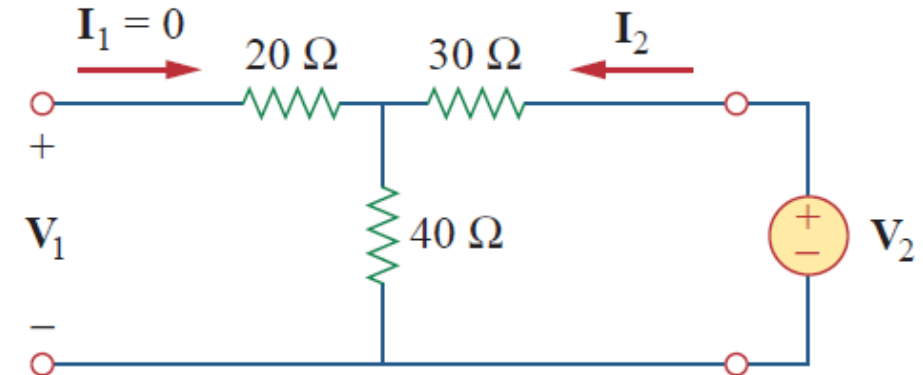
**Solution**

$$V_1 = z_{11}I_1 + z_{12}I_2$$

$$V_2 = z_{21}I_1 + z_{22}I_2$$

$$z_{12} = \left. \frac{V_1}{I_2} \right|_{I_1=0} = \frac{I_2 \cdot 40}{I_2} = 40 \Omega$$

$$z_{22} = \left. \frac{V_2}{I_2} \right|_{I_1=0} = \frac{I_2 \cdot (30 + 40)}{I_2} = 70 \Omega$$



$$\mathbf{Z} = \begin{bmatrix} 60 & 40 \\ 40 & 70 \end{bmatrix} \Omega, \quad z_{12} = z_{21} \rightarrow \text{reciprocal}$$

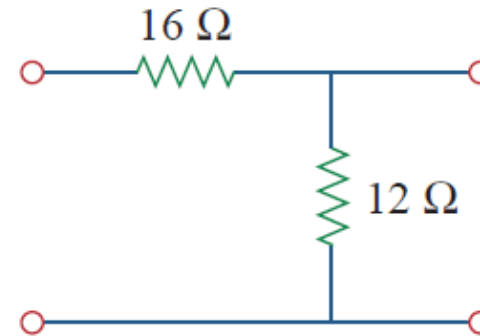
# Z Parameters



Example TWO.02 – Determine the Z parameters of the L-section circuit.

$$V_1 = z_{11}I_1 + z_{12}I_2$$

$$V_2 = z_{21}I_1 + z_{22}I_2$$



**Solution**  $\mathbf{z} = \begin{bmatrix} 28 & 12 \\ 12 & 12 \end{bmatrix} \Omega$ ,  $z_{12} = z_{21} \rightarrow$  *reciprocal*

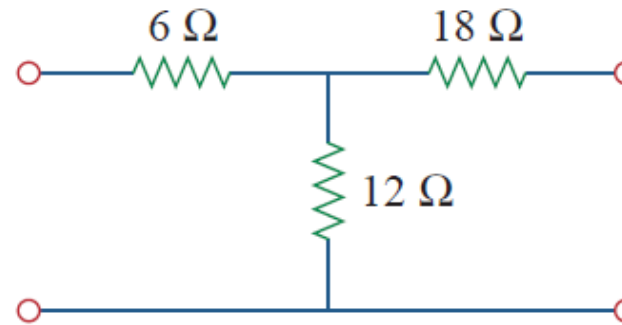
# Y Parameters (SC)



Example TWO.03 – Calculate y parameters of the T-section circuit.

$$I_1 = y_{11}V_1 + y_{12}V_2$$

$$I_2 = y_{21}V_1 + y_{22}V_2$$



**Solution**

$$Y = \begin{bmatrix} 75.77 & -30.3 \\ -30.3 & 45.47 \end{bmatrix} mS, \quad y_{12} = y_{21} \rightarrow \text{reciprocal nw.}$$

# Hybrid Parameters



Example **TWO.04** – Determine the hybrid parameters for the T-section.

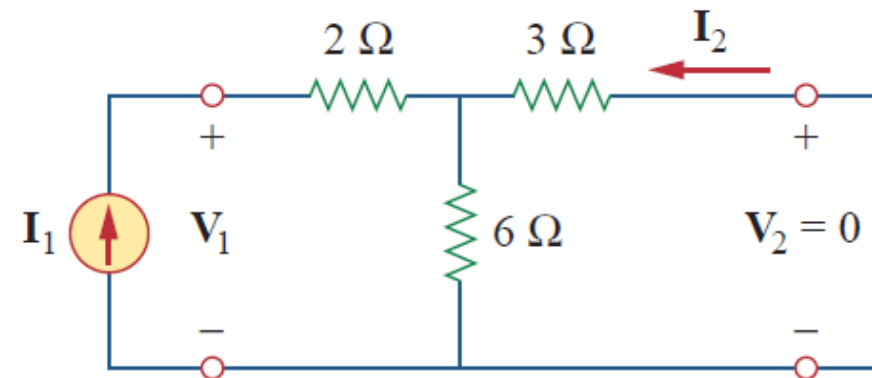
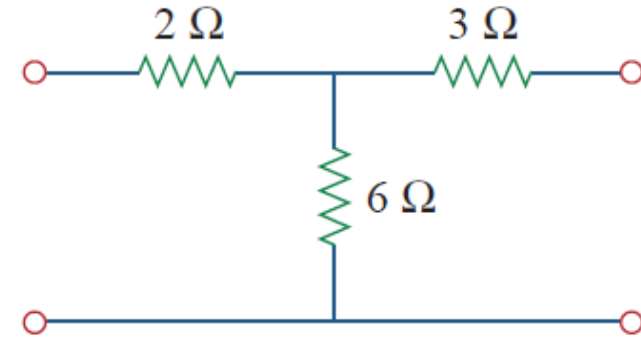
Solution  $V_1 = h_{11}I_1 + h_{12}V_2$        $I_2 = h_{21}I_1 + h_{22}V_2$

**( $h_{11}$ ):**  $V_1 = h_{11}I_1 + h_{12} \cdot 0$

$$h_{11} = \left. \frac{V_1}{I_1} \right|_{V_2=0} = 2 + 3 \times 6 = 4 \Omega$$

**( $h_{21}$ ):**  $-I_2 = \frac{6}{6+3}I_1 = \frac{2}{3}I_1$

$$h_{21} = \left. \frac{I_2}{I_1} \right|_{V_2=0} = -\frac{2}{3}$$



# Hybrid Parameters



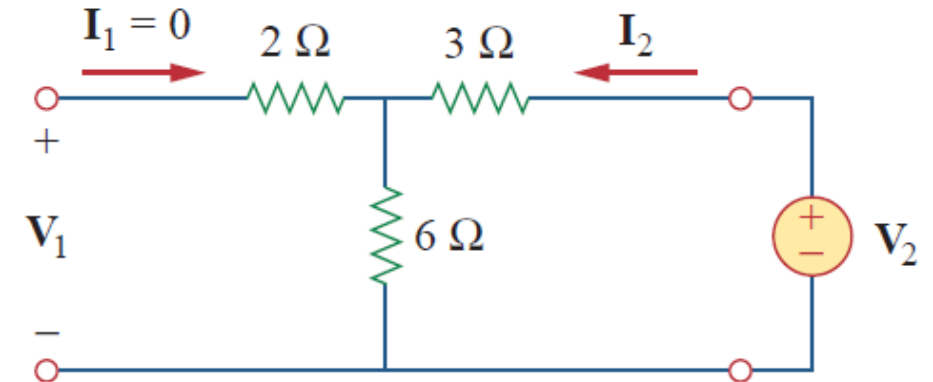
$$V_1 = h_{11}I_1 + h_{12}V_2 \quad I_2 = h_{21}I_1 + h_{22}V_2$$

$$(h_{12}): V_1 = \frac{6}{6+3}V_2 = \frac{2}{3}V_2$$

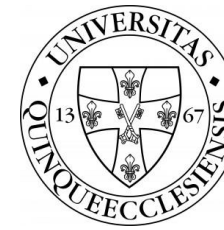
$$h_{12} = \left. \frac{V_1}{V_2} \right|_{I_1=0} = \frac{2}{3}$$

$$(h_{22}): V_2 = (3+6)I_2 = 9I_2$$

$$h_{22} = \left. \frac{I_2}{V_2} \right|_{I_1=0} = \frac{1}{9} \text{ S}$$



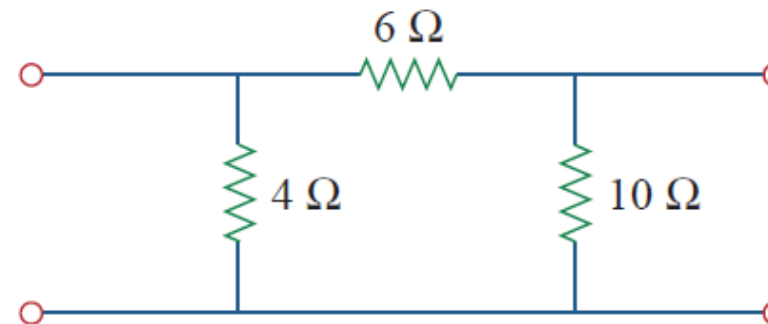
# Hybrid Parameters



Example TWO.05 – Determine the hybrid parameters for the  $\pi$ -section.

$$V_1 = h_{11}I_1 + h_{12}V_2$$

$$I_2 = h_{21}I_1 + h_{22}V_2$$



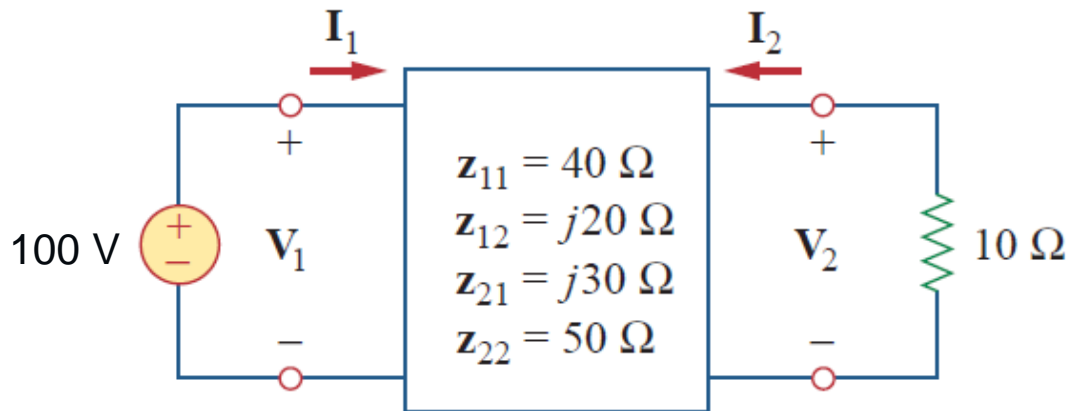
**Solution**

$$H = \begin{bmatrix} 2.4 \Omega & 0.4 \\ -0.4 & 200 \text{ mS} \end{bmatrix}$$

# Network Analysis



Example **TWO.06** – Find  $I_1$  and  $I_2$  in the circuit in Figure.



**Solution** (*non-reciprocal network!*)

$$V_1 = 40I_1 + j20I_2$$

$$V_2 = j30I_1 + 50I_2$$

$$V_1 = 100, V_2 = -10I_2$$

$$(1): 100 = 40I_1 + j20I_2$$

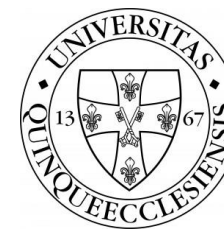
$$(2): -10I_2 = j30I_1 + 50I_2 \rightarrow I_1 = j2I_2$$

$$(2) \rightarrow (1): 100 = j80I_2 + j20I_2 \rightarrow I_2 = -j \text{ A}$$

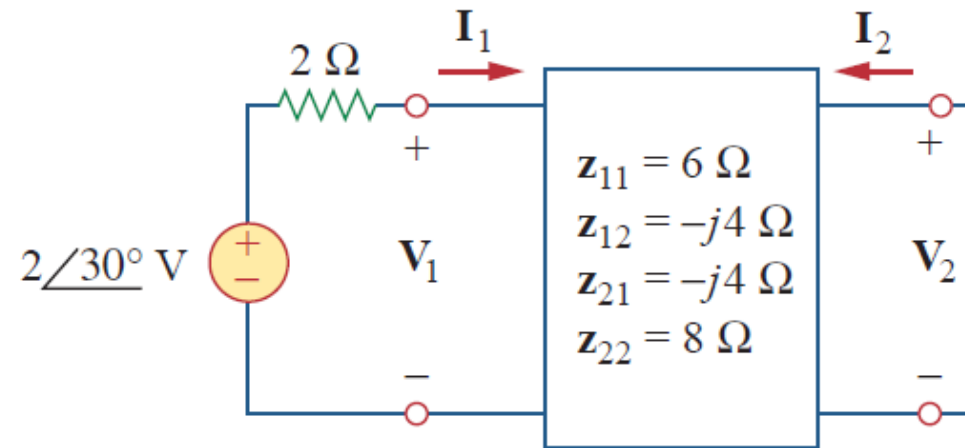
$$I_1 = j2I_2 = 2 \text{ A}$$



# Network Analysis



Example TWO.07 – Find  $I_1$  and  $I_2$  in the circuit in Figure.



**Solution** (*reciprocal network!*)

$$I_1 = 200 e^{j30^\circ} \text{ mA}, \quad I_2 = 100 e^{j120^\circ} \text{ mA}$$

# Network Analysis



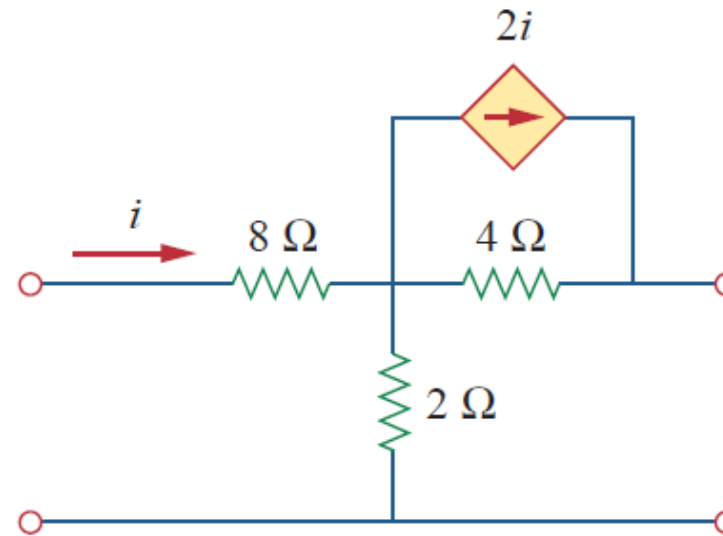
Example **TWO.08** – Determine the Y parameters for the two-port shown in Figure.

Solution  $I_1 = y_{11}V_1 + y_{12}V_2$

$$I_2 = y_{21}V_1 + y_{22}V_2$$

$$y_{11} = \left. \frac{I_1}{V_1} \right|_{V_2=0} \quad y_{12} = \left. \frac{I_1}{V_2} \right|_{V_1=0}$$

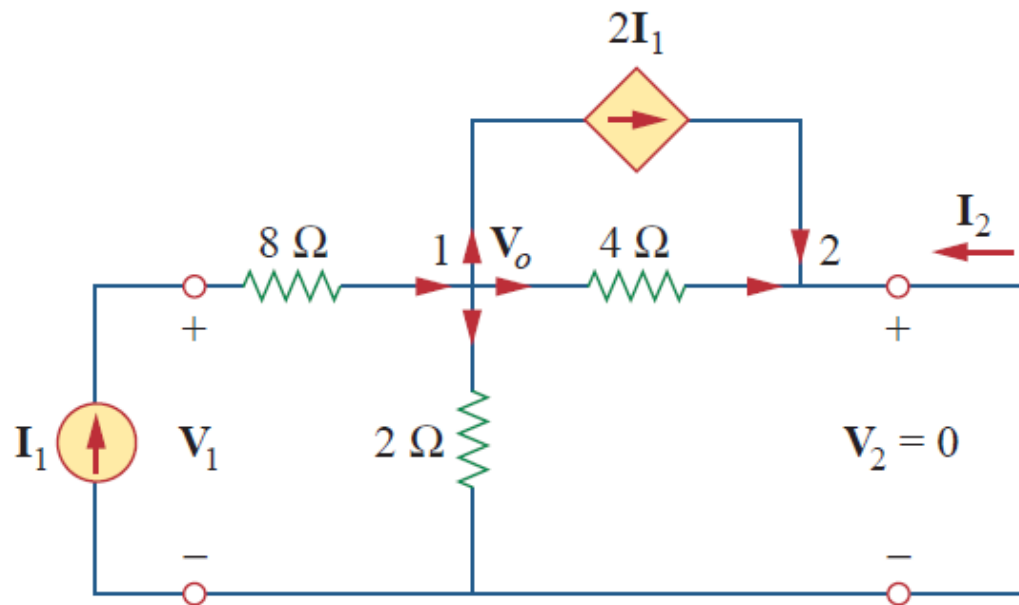
$$y_{21} = \left. \frac{I_2}{V_1} \right|_{V_2=0} \quad y_{22} = \left. \frac{I_2}{V_2} \right|_{V_1=0}$$



# Network Analysis



$$y_{11} = \left. \frac{I_1}{V_1} \right|_{V_2=0}$$



$$(\text{node 1}): \frac{V_1 - V_0}{8} = 2I_1 + \frac{V_0}{2} + \frac{V_0 - 0}{4}$$

$$(\text{but}): I_1 = \frac{V_1 - V_0}{8} \rightarrow \frac{V_1 - V_0}{8} = 2 \frac{V_1 - V_0}{8} + \frac{3V_0}{4}$$

$$\rightarrow 0 = \frac{V_1 - V_0}{8} + \frac{3V_0}{4} \rightarrow$$

$$(1): 0 = V_1 - V_0 + 6V_0 \rightarrow V_1 = -5V_0$$

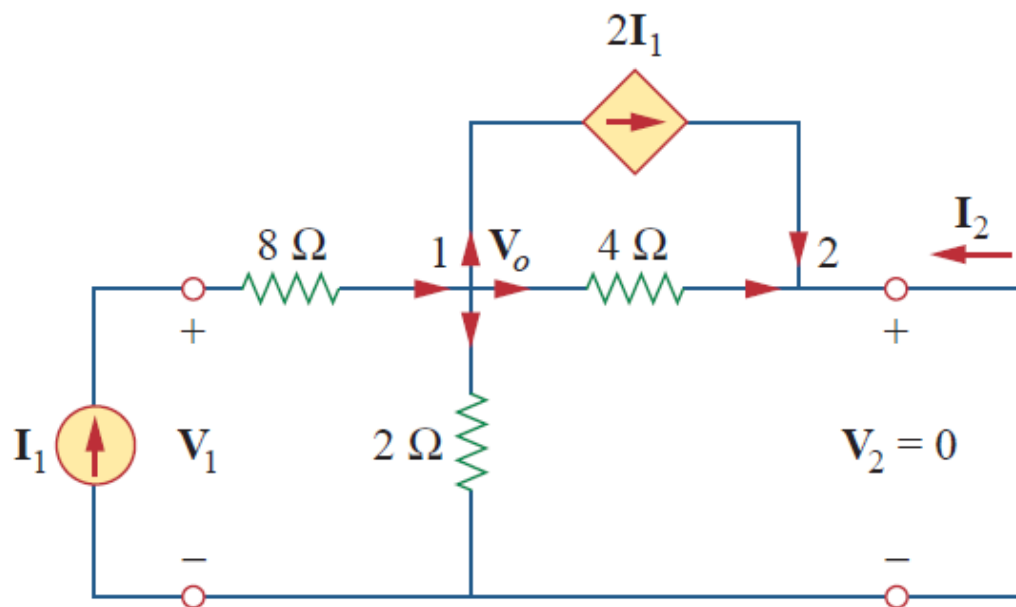
$$(2): I_1 = \frac{V_1 - V_0}{8} = \frac{-5V_0 - V_0}{8} = -0.75V_0$$

$$y_{11} = \frac{I_1}{V_1} = \frac{-0.75V_0}{-5V_0} = 0.15 \text{ S}$$

# Network Analysis



$$y_{21} = \left. \frac{I_2}{V_1} \right|_{V_2=0}$$



$$(1): V_1 = -5V_0$$

$$(2): I_1 = -0.75V_0$$

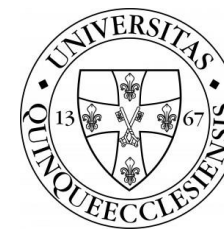
$$(\text{node } 2): \frac{V_0 - 0}{4} + 2I_1 + I_2 = 0$$

$$\frac{V_0}{4} + 2 \cdot (-0.75V_0) = -I_2 = -1.25V_0$$

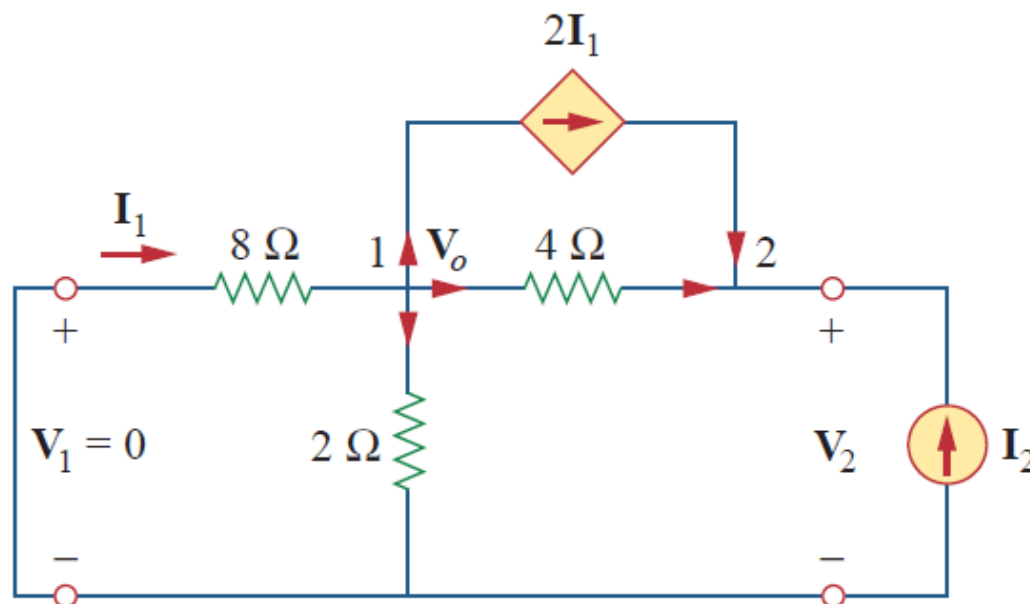
$$y_{21} = \frac{I_2}{V_1} = \frac{1.25V_0}{-5V_0} = -0.25 \text{ S}$$



# Network Analysis



$$y_{22} = \left. \frac{I_2}{V_2} \right|_{V_1=0}$$



(3):  $V_2 = 2.5V_0$       (but):  $I_1 = \frac{0 - V_0}{8}$

(node 2):  $\frac{V_0 - V_2}{4} + 2I_1 + I_2 = 0$

$$-I_2 = \frac{V_0 - (2.5V_0)}{4} - 2 \frac{V_0}{8} = -0.625V_0$$

$$y_{22} = \frac{I_2}{V_2} = \frac{0.625V_0}{2.5V_0} = 0.25 \text{ S}$$

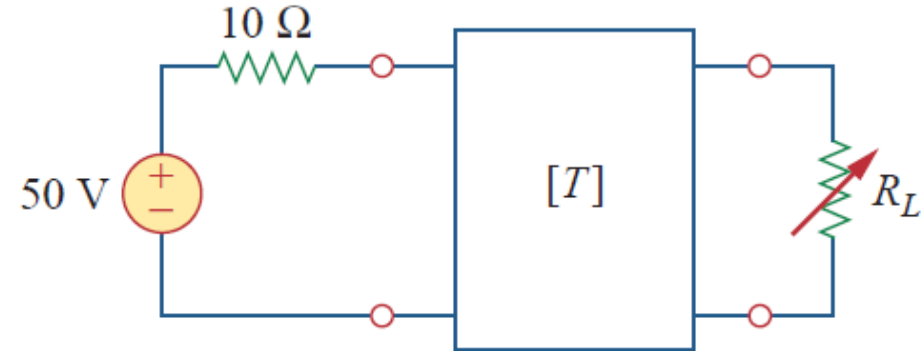
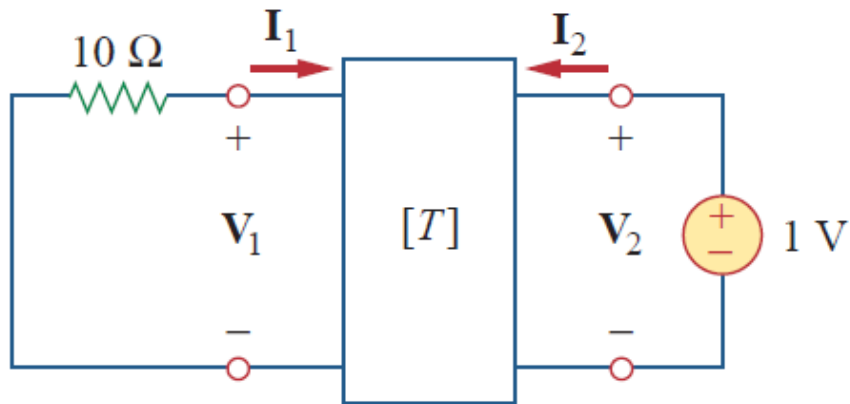
$y_{12} \neq y_{21} \rightarrow$  non reciprocal network

# Network Analysis



**Example TWO.09** – The transmission parameters of the two-port network in Figure are...  $T = \begin{bmatrix} 4 & 20 \Omega \\ 0.1 S & 2 \end{bmatrix}$   
 The output port is connected to a variable load for maximum power transfer. Find  $R_L$  and the maximum power transferred.

**Solution**  $Z_{Th} = ? , V_{Th} = ?$



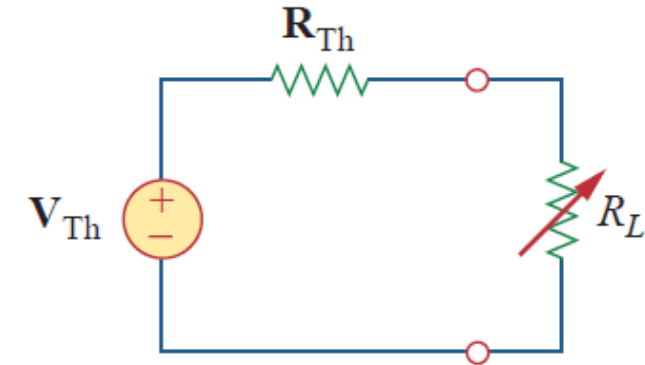
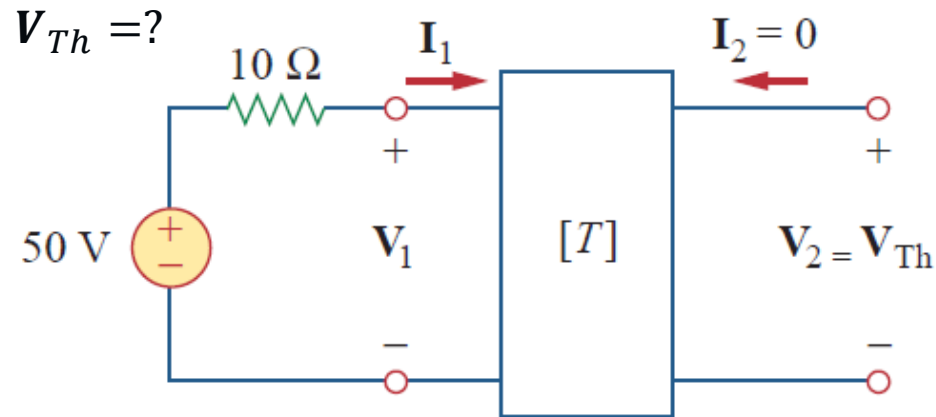
$$(1): V_1 = 4V_2 - 20I_2$$

$$(2): I_1 = 0.1V_2 - 2I_2$$

$$(3): V_1 = -10I_1 \rightarrow (1) \rightarrow -10I_1 = 4V_2 - 20I_2 \rightarrow (4): I_1 = -0.4V_2 + 2I_2$$

$$(2), (4): \rightarrow 0.1V_2 - 2I_2 = -0.4V_2 + 2I_2 \rightarrow 0.5V_2 = 4I_2 \rightarrow Z_{Th} = \frac{V_2}{I_2} = \frac{4}{0.5} = 8 \Omega$$

# Network Analysis



$$(1): V_1 = 4V_2 - 20I_2$$

$$(2): I_1 = 0.1V_2 - 2I_2$$

$$T = \begin{bmatrix} 4 & 20 \Omega \\ 0.1 S & 2 \end{bmatrix}$$

$$(3): I_2 = 0, V_1 = 50 - 10I_1$$

$$(1), (3) \rightarrow (4): 50 - 10I_1 = 4V_2$$

$$(2) \rightarrow (4): 50 - V_2 = 4V_2 \rightarrow V_2 = V_{Th} = 10 V$$

$$R_L = Z_{Th} = R_{Th} = 8 \Omega$$

$$P = I^2 \cdot R_L = \left( \frac{V_{Th}}{2R_L} \right)^2 \cdot R_L =$$

$$= \frac{V_{Th}^2}{4R_L} = \frac{100}{4 \cdot 8} = 3.125 W$$



# Interconnections



Example **TWO.10** – Find the  $y$  parameters of the two-port in Figure.

Solution

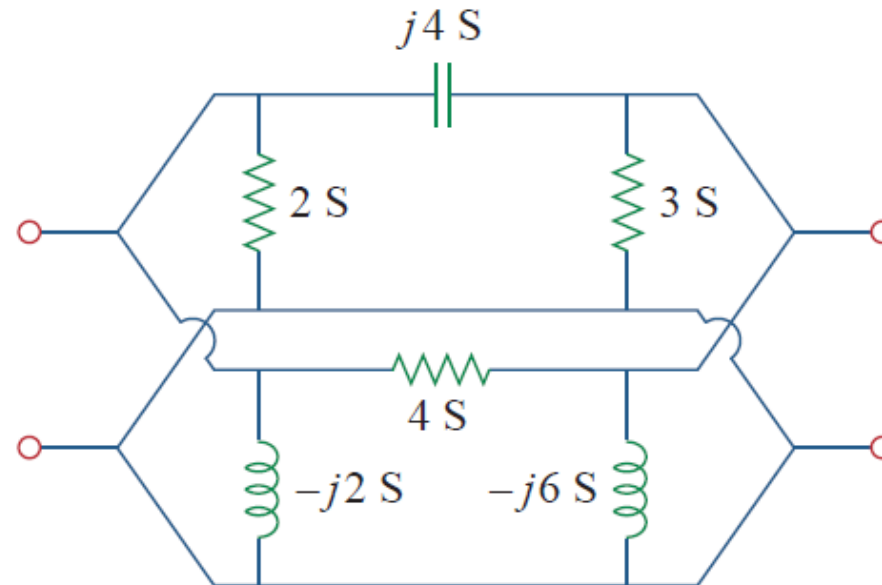
$$I_1 = y_{11}V_1 + y_{12}V_2$$

$$I_2 = y_{21}V_1 + y_{22}V_2$$

$$Y_a = \begin{bmatrix} 2 + j4 & -j4 \\ -j4 & 3 + j4 \end{bmatrix}$$

$$Y_b = \begin{bmatrix} 4 - j2 & -4 \\ -4 & 4 - j6 \end{bmatrix}$$

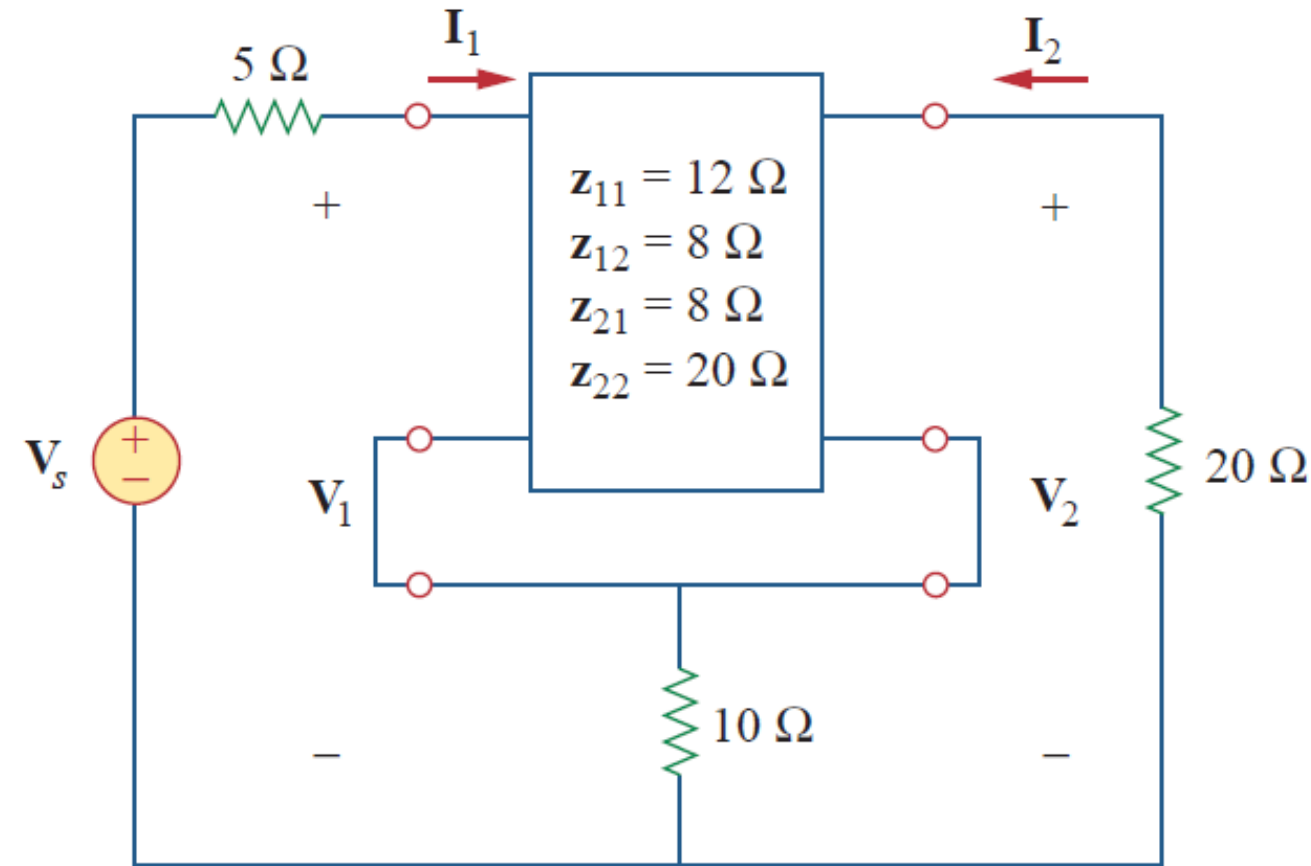
$$Y = Y_a + Y_b = \begin{bmatrix} 2 + j4 & -j4 \\ -j4 & 3 + j4 \end{bmatrix} + \begin{bmatrix} 4 - j2 & -4 \\ -4 & 4 - j6 \end{bmatrix} = \begin{bmatrix} 6 + j2 & -4 - j4 \\ -4 - j4 & 7 - j2 \end{bmatrix}$$



# Interconnections



Example TWO.11 – Evaluate  $V_2/V_s$  in the circuit in Figure.



**Solution**  $z_{12b} = z_{21b} = 10 = z_{11b} = z_{22b}$

$$\mathbf{Z} = \mathbf{Z}_a + \mathbf{Z}_b = \begin{bmatrix} 12 & 8 \\ 8 & 20 \end{bmatrix} + \begin{bmatrix} 10 & 10 \\ 10 & 10 \end{bmatrix} = \begin{bmatrix} 22 & 18 \\ 18 & 30 \end{bmatrix}$$

$$(1): V_1 = z_{11}I_1 + z_{12}I_2 = 22I_1 + 18I_2$$

$$(2): V_2 = z_{12}I_1 + z_{22}I_2 = 18I_1 + 30I_2$$

$$(3): V_1 = V_s - 5I_1 \quad (4): V_2 = -20I_2 \rightarrow I_2 = -\frac{V_2}{20}$$

$$(3), (4) \rightarrow (1): V_s - 5I_1 = 22I_1 - \frac{18}{20}V_2$$

$$\rightarrow (5): V_s = 27I_1 - 0.9V_2$$

$$(4) \rightarrow (2): V_2 = 18I_1 - \frac{30}{20}V_2 \rightarrow (6): I_1 = \frac{2.5}{18}V_2$$

$$(6) \rightarrow (5): V_s = 27 \frac{2.5}{18} V_2 - 0.9V_2 = 2.85V_2 \rightarrow \frac{V_2}{V_s} = \frac{1}{2.85} = 0.35$$

# Wave Impedance



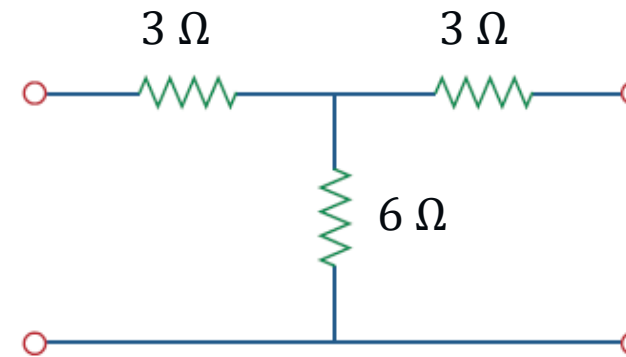
Example **TWO.12** – Calculate the wave impedance of the symmetric T-section.

Solution

$$\mathbf{Z}_{SC} = 3 + 3 \times 6 = 3 + \frac{3 \cdot 6}{3 + 6} = 5 \Omega$$

$$\mathbf{Z}_{OC} = 3 + 6 = 9 \Omega$$

$$\mathbf{Z}_0 = \sqrt{\mathbf{Z}_{SC} \cdot \mathbf{Z}_{OC}} = \sqrt{45} = 3\sqrt{5} \Omega$$

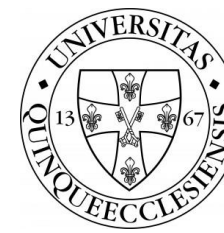


Another way...  $\mathbf{Z}_{IN} = 3 + 6 \times (3 + \mathbf{Z}_0) = \mathbf{Z}_0 \rightarrow 3 + \frac{18 + 6\mathbf{Z}_0}{9 + \mathbf{Z}_0} = \mathbf{Z}_0$

$$27 + 3\mathbf{Z}_0 + 18 + 6\mathbf{Z}_0 = 9\mathbf{Z}_0 + \mathbf{Z}_0^2$$

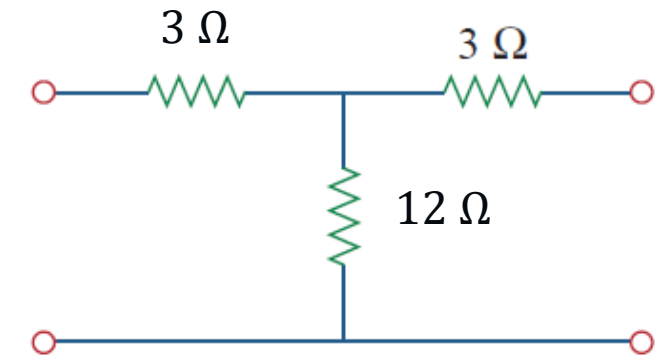
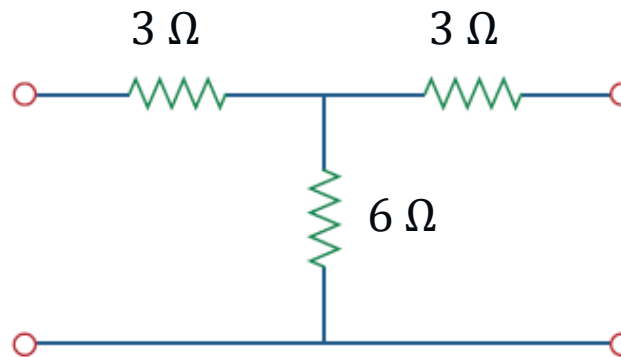
$$45 = \mathbf{Z}_0^2 \rightarrow \mathbf{Z}_0 = \sqrt{45} = 3\sqrt{5} \Omega$$

# Wave Impedance



## Example TWO.13

Calculate wave impedance and Z parameters of the symmetric T-section using Bartlett's theorem.



**Solution** (with 'half-network')

$$\mathbf{Z}_I = \mathbf{Z}_{SC-half} = (\mathbf{Z}_{11} - \mathbf{Z}_{12}) = 3 \Omega$$

$$\mathbf{Z}_{II} = \mathbf{Z}_{OC-half} = (\mathbf{Z}_{11} + \mathbf{Z}_{12}) = 3 + 12 = 15 \Omega$$

$$\mathbf{Z}_0 = \sqrt{\mathbf{Z}_I \cdot \mathbf{Z}_{II}} = \sqrt{3 \cdot 15} = 3\sqrt{5} \Omega \leftarrow \text{refer to the prev. example}$$

$$\mathbf{Z}_{11} = \mathbf{Z}_{22} = \frac{\mathbf{Z}_{II} + \mathbf{Z}_I}{2} = \frac{15 + 3}{2} = 9 \Omega \quad \mathbf{Z}_{12} = \mathbf{Z}_{21} = \frac{\mathbf{Z}_{II} - \mathbf{Z}_I}{2} = \frac{15 - 3}{2} = 6 \Omega$$

# App. – Transmission Line Analysis



## Example TWO.14

Determine the transmission parameters of the  $\Pi$ -equivalent circuit of a long power transmission line operating with a frequency of 50 Hz. Determine if this two-port is reciprocal or symmetrical. ( $U_1 = A_{11}U_2 + A_{12}I_2, I_1 = A_{21}U_2 + A_{22}I_2$ )

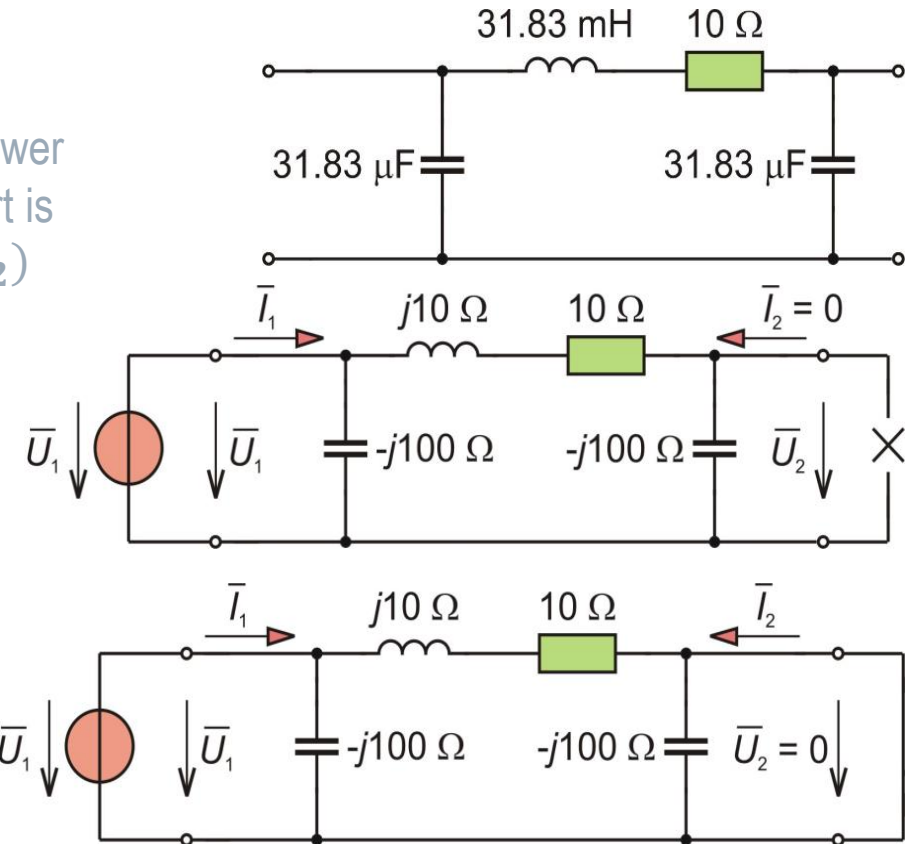
**Solution**

$$X_L = \omega L = 2\pi \cdot 50 \cdot 31.83 \cdot 10^{-3} = 10\Omega$$

$$X_C = \frac{1}{\omega C} = \frac{1}{2\pi \cdot 50 \cdot 31.83 \cdot 10^{-6}} = 100\Omega$$

$$A_{11} = \frac{U_1}{U_2} = \frac{U_1}{U_1 \frac{-j100}{j10 + 10 - j100}} = \frac{10 - j90}{-j100} = 0.9 + j0.1$$

$$A_{12} = \frac{U_1}{I_2} = \frac{U_1}{-I_1 \frac{-j100}{j10 + 10 - j100}} = \frac{U_1}{\frac{U_1}{-j100 \times (j10 + 10)} \cdot \frac{-j100}{j10 + 10 - j100}} = \dots = (-10 - j10)\Omega$$



# App. – Transmission Line Analysis



$$A_{21} = \frac{I_1}{U_2} = \frac{I_1}{I_1 \frac{-j100}{-j100 + j10 + 10 - j100} (-j100)} =$$

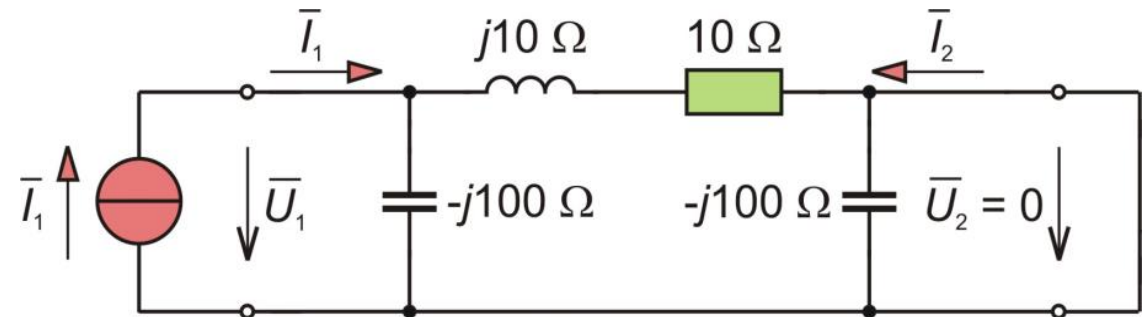
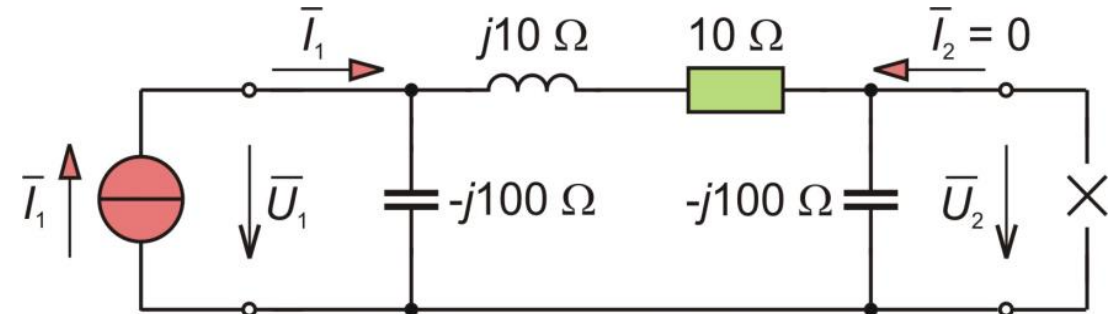
$$= \frac{10 - j190}{10000} = (-0.001 + j0.019) S$$

$$A_{22} = \frac{I_1}{I_2} = \frac{I_1}{-I_1 \frac{-j100}{-j100 + j10 + 10}} = \frac{-j100 + j10 + 10}{j100} = -0.9 + j0.1$$

$$A = \begin{bmatrix} 0.9 + j0.1 & (-10 - j10)\Omega \\ (-0.001 + j0.019) S & -0.9 - j0.1 \end{bmatrix}$$

$$\Delta A = A_{11}A_{22} - A_{12}A_{21} = \dots = -1$$

$$A_{11} = -A_{22} \left. \vphantom{\Delta A} \right\} \rightarrow \text{reciprocal \& symmetrical}$$



# App. – Transmission Line Analysis



## Example TWO.15

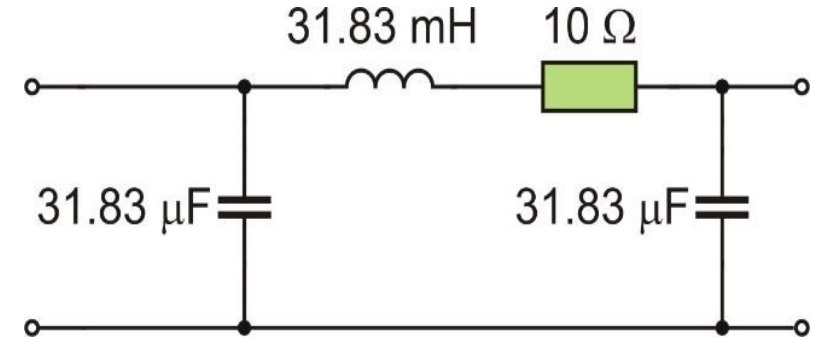
Calculate the wave impedance of the  $\Pi$ -equivalent network of a long power transmission line operating with a frequency of 50 Hz.

### Solution

$$Z_I = -jX_C = -j \frac{1}{\omega C} = -j100 = 100 e^{-j90^\circ} \Omega$$

$$Z_{II} = \left( \frac{R}{2} + j \frac{X_L}{2} \right) \times (-jX_C) = (5 + j5) \times (-j100) = \frac{\sqrt{50} e^{j45^\circ} 100 e^{-j90^\circ}}{5 - j95} = \frac{707.1 e^{-j45^\circ}}{95.13 e^{-j86.99^\circ}} = 7.43 e^{j41.99^\circ} \Omega$$

$$Z_0 = \sqrt{Z_I Z_{II}} = \sqrt{100 e^{-j90^\circ} 7.43 e^{j41.99^\circ}} = \sqrt{743 e^{-j48.01^\circ}} = 27.26 e^{-j24^\circ} = 24.89 - j10.90 \Omega$$



# Questions

